A Session Hijacking Attack against a Device-Assisted Physical-Layer Key Agreement

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Abstract—Physical-layer key agreement is used to generate a shared key between devices on demand. Such schemes utilize the characteristics of the wireless channel to generate the shared key from the device-to-device channel. As all characteristics are time-dependent and location-dependent, it is hard for eavesdroppers to get the key. However, most researches in this area use passive attack models whereas active attacks that aim at manipulating the channel and key are also possible. PHY-UIR (PHYsical layer key agreement with User Introduced Randomness) is a solution similar to the Diffie-Hellman protocol against such a kind of active attack. The users (devices) introduce their own randomness to help to prevent active attacks. In this paper, we analyze the possibility of launching a session hijacking attack on PHY-UIR to allow an attacker to control the shared key established. The session hijacking attack manipulates the key agreement through a man-in-the-middle (MITM) interaction and forces legitimate devices to run the PHY-UIR protocol with the attacker. Our simulation and experiment results validate our attack and show the high performance of our attack on manipulating the generated key. We also propose PHY-UIR+ where devices simultaneously exchange information about the established shared keys, which allows them to detect whether they have agreed to different keys with a third party.

Index Terms—Physical-layer security, key management.

I. INTRODUCTION

THE Internet-of-Things (IoT) is rapidly growing and will continue to do so with the help of other emerging technologies, like 5th-generation (5G) wireless systems. For the IoT to really reach its envisaged goal of interconnecting all "things" we must be able to ensure secure device-to-device communication. To provide basic security services between devices, these devices must share cryptographic keys. Without effective key management within a system, other security services related to confidentiality, integrity and authentication cannot function correctly as these all rely on devices holding suitable cryptographic keys. Key management is not easy, and becomes increasingly difficult as systems scale, e.g. each device needs to share a unique secret symmetric key with every other device it might wish to communicate with. This leads to the storing of a large number of keys to ensure that two devices that do meet already have a shared key.

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Key management, especially key establishment (the process whereby two devices end up with a shared key) is therefore important yet challenging [1], [2], [3], [4], [5].

Physical layer key generation schemes [6], [7] are a kind of key establishment method that generates the shared key based on the physical properties of channel measurements of two devices and some schemes have been proven to be information-theoretic secure [8]. If considering current key management standards, like ISO/IEC 11770, models herein always have the assumption that the parties to the establishment already have a trusted relationship, i.e. a shared key, and that they can then establish a new key. The low complexity of generating physical-layer keys on demand and the benefit that it needs no prior shared secrets make this type of key establishment a promising alternative to classical cryptographic methods of key establishment [9]. Studying the vulnerabilities of physical layer key generation schemes to construct more robust and secure schemes is one of the important future research scopes[10], while the approach has also been proposed in industrial applications [11].

While passive attacks on physical layer key generation schemes attract much attention, only a few authors focus on analyzing active attacks. There are two kinds of active attacks. One is the jamming attack [7], [12], which aims at decreasing the speed of key generation. The other is the key manipulation attack [13], [14], which aims at manipulating parts of or the whole generated key. To mitigate the second kind of active attack, researchers mainly focus on authenticating transmitters by leveraging the fingerprints of previous signals that have not been attacked. There are two kinds of fingerprints: hardware fingerprints [15], [16] and channel fingerprints [17], [18]. However, hardware fingerprints have been broken by estimating and reproducing attacks [19], while channel fingerprints are vulnerable to an analog man-in-the-middle (MITM) attack by contaminating the signals used to generate fingerprints [20].

PHY-UIR [14] is a solution mitigating key manipulation attacks, working on similar principles to the Diffie-Hellman key exchange protocol. The question is whether this protocol is also vulnerable to the MITM attack, which the Diffie-Hellman protocol suffers from. In our previous paper [21], we proposed the possibility of launching a kind of MITM attack called the session hijacking attack, which utilizes a fast reactive jamming technology to force legitimate devices generating the shared key with signals from the attacker when these devices are executing the PHY-UIR protocol. This paper adds theoretical analysis on the factors that impact on the success rate of the session hijacking attack and proposes a countermeasure
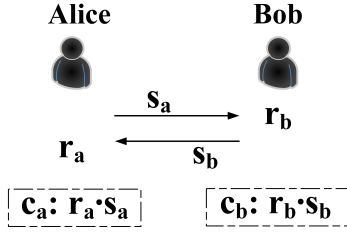
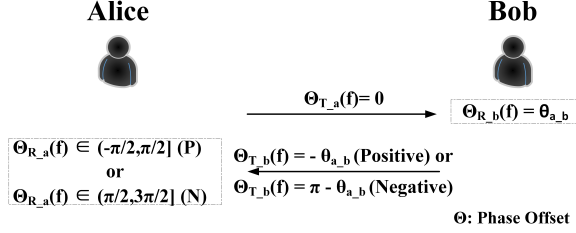Fig. 1: RSS quantization procedure in PHY-UIR

Fig. 2: Phase quantization procedure in PHY-UIR

against the session hijacking attack.

The main contributions of our paper are as follows:

- We analyze the feasibility of the session hijacking attack in more detail and identify the main factors that impact on the success rate of the attack.
- We investigate the impact of these factors by simulation, and then conduct experiments with software-defined radios to validate the simulation result.
- An enhanced scheme PHY-UIR$^+$ is proposed and tested to mitigate the attack in question.

II. PHY-UIR KEY GENERATION METHOD

In this section, we depict the workflow of the PHY-UIR key generation method. As with other generation schemes, both users exchange orthogonal frequency-division multiplexing (OFDM) signals and generate the key from received OFDM signals, but PHY-UIR extracts the key from two aspects: received signal strength (RSS) and phase.

PHY-UIR extracts two bits from each round of signal exchange. We show the process in Figure 1. Unlike with other key agreement schemes, devices in PHY-UIR transmit different signals in each round and multiply what they receive with what they transmit to generate the shared signals and apply a multi-level quantization method proposed in [22]. PHY-UIR measures the phase of the subcarrier frequency of the signal to determine the sign of the corresponding two bits. We illustrate the process in Figure 2. Bob receives the signal with a phase offset θ_{a_b}. Then he compensates for this phase offset by adjusting the phase offset of its signal. Adding an additional π phase change indicates a negative sign, otherwise, the value has a positive sign. Alice can also know the sign from the phase offset she calculates from what she receives.

III. ANALYSIS OF SESSION HIJACKING ATTACK

In this section, we depict our session hijacking attack in detail. Our attacker hijacks the communication during the channel probing period and establishes two different shared

keys with each of the two devices separately. Our attacker can act as a legitimate participant to communicate with both devices. The main element of the attack is to eliminate the randomness introduced by the user and replace the randomness of channel state information (CSI). To achieve this, the attacker transmits powerful fake signals to overshadow legitimate signals. In this way, the randomness of the fake signals and the CSI between the attacker and users are known to the attacker and the attacker can generate shared keys with Alice or Bob. The shared keys consist of two parts, which are generated by RSS and phase quantization schemes. We analyze how our attack affects these two schemes separately. Table I summarizes the notations we use in this paper.

A. OFDM Model

First we discuss background information of OFDM. In a discrete-time system, an OFDM symbol consists of many sample points that can be expressed as

$$s(k) = \frac{1}{\sqrt{N}} \sum_{n=0}^{N-1} S(n) e^{j2\pi nk/N} \qquad (1)$$

where $s(k)$ is the value of the kth point of the OFDM symbol, N is the number of subcarriers and $S(n)$ is the data that has been modulated onto the nth subcarrier in this symbol. Here we assume that the original phase offset is 0.

Parts of an OFDM symbol that corresponds to the nth subcarrier can be expressed as

$$s(k,n) = \frac{1}{\sqrt{N}} S(n) e^{j2\pi nk/N} \qquad (2)$$

After transmission, the OFDM symbol will experience various reflections, backscattering and refraction in the multipath channel, which we represent by adopting the multipath model in [23]. In each channel tap, the channel impulse response (CIR) can be expressed as

$$h(l,t) = \mu(\tau_l, t) e^{j\phi(\tau_l, t)} \qquad (3)$$

where l represents the lth channel path. $\mu_l(t)$ and $\phi_l(t)$ are the amplitude attenuation and phase offset caused by this channel path at time t. The total CIR of this channel can be written as

$$h_{whole}(\tau, t) = \sum_{l=0}^{L-1} h(l,t)\delta(\tau - \tau_l) \qquad (4)$$

where L is the total number of the channel taps, τ_l is the delay time of the lth channel tap and δ is the Dirac delta function. As in the model, the tap delays are assumed to be sample-spaced, which means that $\tau_l = lT$. T is the sampling period of the receiving device.

According to [24], a channel in a rich scattering environment is divided into small time frames. Each time frame can be treated as a wide sense stationary (WSS) random process which means each time frame is uncorrelated with the other one. Each attacking unit in our paper can be treated as a WSS random process. As an example, we analyze our attack in one attack unit. For coherence time T_{co}, the CIRs of the channels are constant and the length of one attack unit is less than one coherence time, for simplicity we omit t in equations 3 and 4.

TABLE I: Symbols for equations used in this paper

Symbol	Notation
A_a, A_b A_{att}	The coefficients of signals from Alice, Bob and the attacker
$s(k)$	The value of the kth point of an OFDM symbol
$s(k,n)$	The value of the kth point of an OFDM symbol at the nth subcarrier
$r(k,n)$	The value of the kth point of the signal received at the receiver corresponding to the nth carrier
$s_a(k,n)$ $(s_b(k,n))$	Signals transmitted by Alice (Bob)
$s_{att\_a}(k,n)$ $(s_{att\_b}(k,n))$	Signals transmitted by the attacker to Alice (Bob)
$s_{att\_att}(k,n)$	Signals transmitted by the attacker to the attacker
$r_a(k,n)$ $(r_b(k,n))$	Signals received by Alice (Bob)
$r_{att\_a}(k,n)$ $(r_{att\_b}(k,n))$	Signals received by the attacker from Alice (Bob)
$c_a(k,n)$ $(c_b(k,n))$	Final signals used to generate the shared key by Alice (Bob)
$c_{a\_att}(k,n)$ $(c_{b\_att}(k,n))$	Final signals used to generate the shared key by the attacker corresponding to Alice (Bob)
$K\_S_a(k,n)$	Main parts of $c_{a\_att}(k,n)$ to generate the shared key
$w(k,n)$	White Gaussian Noise
$h_{whole}(\tau,t)$	The total CIR of a channel
$h(l,t)$	CIR of the lth channel tap
$h_{a\_b}(l_{a\_b})$	CIR of the $l_{a\_b}^{th}$ channel tap in the channel between Alice and Bob
$h_{a\_att}(l_{a\_att})$	CIR of the $l_{a\_att}^{th}$ channel tap in the channel between Alice and the attacker
$h_{b\_att}(l_{b\_att})$	CIR of the $l_{b\_att}^{th}$ channel tap in the channel between Bob and the attacker
h_{att_att}	CIR of the channel between two antennas on the attacker
τ_l	The delay time of lth channel tap
ε	The time offset between two devices
$\Theta_{R\_b}(n)$ $(\Theta_{R\_att}(n))$	Phase offset measured by Bob (the attacker)
$rand_b(n)$ $(rand_{att}(n))$	A random number generated by Bob (the attacker). It has only two possible values, 0 and 1
k_{b_a} (k_{att_a})	Total offset measured by Alice from the received signals coming from Bob (the attacker)
k_{a_b} (k_{att_b})	Total offset measured by Bob from the received signals coming from Alice (the attacker)
k_{a_att} (k_{b_att})	Total offset measured by the attacker from the received signals coming from Alice (Bob)
$\tau_{l_{a\_b}}$	The delay time of lth channel tap in the channel between Alice and Bob
$\tau_{l_{a\_att}}$	The delay time of lth channel tap in the channel between Alice and the attacker
$\tau_{l_{b\_att}}$	The delay time of lth channel tap in the channel between Bob and the attacker
ε_{b_a} (ε_{a_b})	Time offset between Alice and Bob which is measured by Alice (Bob)
ε_{att_a} (ε_{a_att})	Time offset between Alice and the attacker which is measured by Alice (the attacker)
ε_{att_b} (ε_{b_att})	Time offset between Bob and the attacker which is measured by Bob (the attacker)
τ_{ahead_a} (τ_{ahead_b})	Differences of arrival times between the attack signals and the signals from Alice (Bob).
R,S,H,Z	Equivalent frequency domain value of r,s,h,w
η_{att}, η_b	Amplitude attenuation

In this paper, we also assume that the transmitter and the receiver both have a perfect sampling clock [23]. Also as in a slow fading environment, h can be treated as constant for one OFDM symbol. Then the signal received at the receiver
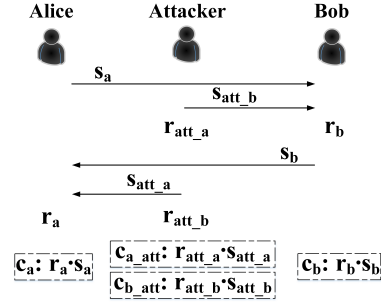

Fig. 3: Workflow of a session hijacking attack on the RSS quantization scheme

corresponding to the nth carrier can be represented by the following equation[23]:

$$r(k,n) = \sum_{l=0}^{L-1} h(l)s(k-\tau_l-\varepsilon,n) + w(k,n) \qquad (5)$$

where ε is the time offset that is caused by the imperfect synchronization. The synchronization is used to find the start point of the signal to extract the data perfectly. Due to the impact of the channel, the noise and the synchronization protocol, there is always imperfect synchronization. w is White Gaussian Noise (WGN).

B. Attack Model

There are three entities in our attack model: an attacker and two legitimate devices, Alice and Bob. Devices generate the shared key with the help of PHY-UIR. The attacker can sniff legitimate signals and transmit attack signals during the channel probing stage to hijack the communication. What the attacker does not know are the amplitudes of the signals before they are transmitted and the channel state between two legitimate devices. The attacker has knowledge of the procedures of PHY-UIR and some settings only used for communication, such as carrier frequency and modulation scheme. We adopt the same wireless network settings used in [25] and during the period of coherence time, all channels between the three entities have constant states.

We analyze the impact of our session hijacking attack on these two schemes below.

C. Hijacking Attack on RSS Quantization Scheme

We illustrate the workflow of our attack on the RSS quantization step in Figure 3. We choose one bidirectional signal exchange as an example. In this attack, s_a and s_b are normal signals with random amplitude. s_{att_b} and s_{att_a} are the attack signals transmitted by the attacker with constant amplitude to Bob and Alice separately. According to Figure 2, Bob will compensate the offset in s_b. Also, the attacker will copy this action to enable the attack that we will depict later in this section. The RSS quantization scheme chooses a certain subcarrier to generate the shared key. The signals at this subcarrier can be expressed as

$$s_a(k,n) = A_a s(k,n) \qquad (6)$$

$$s_b(k,n) = A_b \frac{1}{\sqrt{N}} S(n) e^{j(2\pi nk/N - \Theta_{R\_b}(n) + rand_b*\pi)} \qquad (7)$$

$$s_{att\_b}(k,n) = A_{att} s(k,n) \qquad (8)$$

$$s_{att\_a}(k,n) = A_{att} \frac{1}{\sqrt{N}} S(n) e^{j(2\pi nk/N - \Theta_{R\_att}(n) + rand_{att}*\pi)} \qquad (9)$$

In equation 6 and 7, A_a and A_b are coefficients used by Alice and Bob to change the amplitude of the signals. A_{att} is the coefficients used by the attacker. Each signal has a constant coefficient but from the view of the whole channel probing procedure, A_a and A_b vary randomly from one signal to another to form the randomness introduced by the user while A_{att} is a constant value. Θ_{R_b} is the phase offset measured by Bob. Θ_{R_att} is the phase offset measured by the attacker from the signal transmitted by Alice. Bob and the attacker make a decision on the sign of current signal by $rand$, 0 means positive and 1 means negative. Then the received signals, r_a and r_b, at the certain subcarrier are:

$$r_a(k,n) = \sum_{l_{a\_b}=0}^{L_{a\_b}-1} h_{a\_b}(l_{a\_b})s_b(k_{b\_a},n) \tag{10}$$
$$+ \sum_{l_{a\_att}=0}^{L_{a\_att}-1} h_{a\_att}(l_{a\_att})s_{att\_a}(k_{att\_a},n) + w_a(k,n)$$

where

$$k_{b\_a} = k - \tau_{l_{a\_b}} - \varepsilon_{b\_a} + \tau_{ahead\_a}$$
$$k_{att\_a} = k - \tau_{l_{a\_att}} - \varepsilon_{att\_a}$$

and

$$r_b(k,n) = \sum_{l_{a\_b}=0}^{L_{a\_b}-1} h_{a\_b}(l_{a\_b})s_a(k_{a\_b},n) \tag{11}$$
$$+ \sum_{l_{b\_att}=0}^{L_{b\_att}-1} h_{b\_att}(l_{b\_att})s_{att\_b}(k_{att\_b},n) + w_b(k,n)$$

where

$$k_{a\_b} = k - \tau_{l_{a\_b}} - \varepsilon_{a\_b} + \tau_{ahead\_b}$$
$$k_{att\_b} = k - \tau_{l_{b\_att}} - \varepsilon_{att\_b} \tag{12}$$

In the above two equations, L_{a_b}, L_{a_att} and L_{b_att} are the total number of the channel taps in channels Alice-Bob, Alice-Attacker and Bob-Attacker respectively. The received s_a and s_b also have the same time offset with their relayed versions, because the receiver has synchronized with their relay versions. τ_{ahead_a} and τ_{ahead_b} are the time differences that the signals arrives at the receiver ahead of the attack signals.

According to PHY-UIR, the final signals used to generate parts of the shared key by the RSS quantization scheme at legitimate devices are:

$$c_a(k,n) = \sum_{l_{a\_b}=0}^{L_{a\_b}-1} h_{a\_b}(l_{a\_b})s_a(k,n)s_b(k_{b\_a},n) + s_a(k,n)w_a(k,n)$$
$$+ \sum_{l_{a\_att}=0}^{L_{a\_att}-1} h_{a\_att}(l_{a\_att})s_a(k,n)s_{att\_a}(k_{att\_a},n) \tag{13}$$

and

$$c_b(k,n) = \sum_{l_{a\_b}=0}^{L_{a\_b}-1} h_{a\_b}(l_{a\_b})s_b(k,n)s_a(k_{a\_b},n) + s_b(k,n)w_b(k,n)$$
$$+ \sum_{l_{b\_att}=0}^{L_{b\_att}-1} h_{b\_att}(l_{b\_att})s_b(k,n)s_{att\_b}(k_{att\_b},n) \tag{14}$$

From the view of the attacker, received signals at corresponding subcarrier during the signal transmission time of Alice and Bob are r_{att_a} and r_{att_b} are:

$$r_{att\_a}(k,n) = \sum_{l_{a\_att}=0}^{L_{a\_att}-1} h_{a\_att}(l_{a\_att})s_a(k_{a\_att},n)$$
$$+ h_{att\_att}s_{att\_att}(k,n) + w_{a\_att}(k,n) \tag{15}$$

where

$$k_{a\_att} = k - \tau_{l_{a\_att}} - \varepsilon_{a\_att}$$

and

$$r_{att\_b}(k,n) = \sum_{l_{b\_att}=0}^{L_{b\_att}-1} h_{b\_att}(l_{b\_att})s_b(k_{b\_att},n)$$
$$+ h_{att\_att}s_{att\_att}(k,n) + w_{b\_att}(k,n) \tag{16}$$

where

$$k_{b\_att} = k - \tau_{l_{b\_att}} - \varepsilon_{b\_att}$$

In the above two equations, r_{att_a} and r_{att_b} represent the signal corresponding to the certain subcarrier received by the attacker. These two equations are a little different than equations 10 and 11. As the attack signals are transmitted and received all by the attacker and the receiving antenna and the transmitting antenna are very close at the attacker there is almost no multipath effect. We use h_{att_att} to represent the CIR of the channel between the receiving antenna and the transmitting antenna of the attacker.

As self-interference cancellation technology can be used to eliminate the interference caused by $h_{att\_att}s_{att\_att}$, we adopt the technology in [26], which can reduce the self-interference to the noise floor. This means $h_{att\_att}s_{att\_att}$ in equations 15 and 16 can be merged into w_{a_att} and w_{b_att}.

The final signals c_{a_att} and c_{b_att} that are used for RSS quantization with devices Alice and Bob are:

$$c_{a\_att}(k,n) = \sum_{l_{a\_att}=0}^{L_{a\_att}-1} h_{a\_att}(l_{a\_att})s_a(k_{a\_att},n)s_{att\_a}(k,n)$$
$$+ w_{a\_att}(k,n)s_{att\_a}(k,n) \tag{17}$$

and

$$c_{b\_att}(k,n) = \sum_{l_{b\_att}=0}^{L_{b\_att}-1} h_{b\_att}(l_{b\_att})s_b(k_{b\_att},n)s_{att\_b}(k,n)$$
$$+ w_{b\_att}(k,n)s_{att\_b}(k,n) \tag{18}$$

Now legitimate devices and the attack device all get the signals used for key generation through the RSS quantization scheme. This scheme is performed at a certain subcarrier. Parts of c_a and c_{a_att} that belongs to the certain subcarrier are used to generate the key between Alice and the attacker while corresponding parts of c_b and c_{b_att} are used to generate the key between Bob and the attacker. As Alice and Bob generate the key with the attacker in the same way, we analyze Alice and the attacker only. As the noise can be neglected, the key generated by the attacker to communicate with Alice comes from the rest of c_{a_att} which we call it key source K_S_a.

$$K\_S_a(k,n) = \sum_{l_{a\_att}=0}^{L_{a\_att}-1} h_{a\_att}(l_{a\_att})s_a(k_{a\_att},n)s_{att\_a}(k,n)$$
$$= \sum_{l_{a\_att}=0}^{L_{a\_att}-1} h_{a\_att}(l_{a\_att})A_a\frac{1}{\sqrt{N}}S(n)e^{j2\pi nk_{a\_att}/N}$$
$$\cdot (A_{att}\frac{1}{\sqrt{N}}S(n)e^{j(2\pi nk/N-\Theta_{R\_att}+rand_b*\pi)})$$
$$= \sum_{l_{a\_att}=0}^{L_{a\_att}-1} h_{a\_att}(l_{a\_att})s_a(k,n)s_{att\_a}(k_{att\_a}+k_d,n) \tag{19}$$

where

$$k_d = \varepsilon_{att\_a} - \varepsilon_{a\_att}$$

Alice generates her key based on the power of the signal in equation 13 and the attacker generates corresponding key
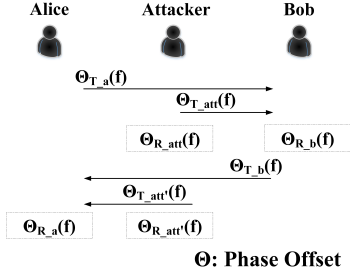
Fig. 4: Workflow of session hijacking attack on phase quantization

based on the power of the signal in equation 19. Let's compare these two equations. We can see that if k_d is equal to zero, K_S_a is equal to parts of c_a. In equation 13, except the parts which is equal to K_S_a and the neglible noise, only $\sum_{l_{a\_b}=0}^{L_{a\_b}-1} h_{a\_b}(l_{a\_b})s_a(k,n)s_b(k_{b\_a},n)$ remains, which has negative impact on our attack as it may change the generated key. Comparing this remaining part in equation 13 with equation 19, we can see that if the power of the signal $\sum_{l_{a\_att}=0}^{L_{a\_att}-1} h_{a\_att}(l_{a\_att})s_{att\_a}(k_{att\_a}+k_d,n)$ has more power than $\sum_{l_{a\_b}=0}^{L_{a\_b}-1} h_{a\_b}(l_{a\_b})s_b(k_{b\_a},n)$, the generated key at Alice should be more similar to the key generated at the attacker. $\sum_{l_{a\_att}=0}^{L_{a\_att}-1} h_{a\_att}(l_{a\_att})s_{att\_a}(k_{att\_a}+k_d,n)$ is the signal received by Alice which is transmitted by the attacker. $\sum_{l_{a\_b}=0}^{L_{a\_b}-1} h_{a\_b}(l_{a\_b})s_b(k_{b\_a},n)$ is the signal received by Alice which is transmitted by the Bob. So there are three factors that have an impact on the phase of the received signals: $\varepsilon_{att\_a}(\varepsilon_{att\_b})$, $\varepsilon_{a\_att}(\varepsilon_{b\_att})$ and the power ratio of the attack signal to the legitimate signal at n^{th} subcarrier at Alice (Bob).

D. Session Hijacking Attack on Phase Quantization Scheme:

After RSS quantization, we depict how the session attack affects the phase quantization. The workflow is shown in Figure 4. At first, Alice transmits a normal signal with no phase offset ($\Theta_{T\_a}(f) = 0$). When the signal arrives at the attacker, it can be written as:

$$R_{att\_a}(n) = \frac{1}{\sqrt{N}}\sum_{k=0}^{N-1} r_{att\_a}(k,n)e^{-j2\pi nk/N}$$
$$= S_a(n)H_{a\_att}(n)e^{-j2\pi k\varepsilon_{a\_att}/N} + Z_{a\_att}(n) \quad (20)$$

where

$$S_a(n) = \frac{1}{\sqrt{N}}\sum_{k=0}^{N-1} s_a(k,n)e^{-j2\pi nk/N}$$
$$Z_{a\_att}(n) = \sum_{k=0}^{N-1} w_{a\_att}(k)e^{-j2\pi nk/N} \quad (21)$$
$$H_{a\_att}(n) = \sum_{l_{a\_att}=0}^{L_{a\_att}-1} h_{a\_att}(l_{a\_att})e^{-j2\pi nl_{a\_att}/N}$$

The attacker transforms $R_{att\_a}(n)$ into a signal s_{R_att} depending on equation 22. Then it extracts the phase offset Θ_{R_att} from s_{R_att} as the attacker knows s. $R_{att\_a}(n)$ can be written as equation 23. η_{att} is the amplitude attenuation.

$$S_{R_{att\_a}}(k,n) = \frac{1}{\sqrt{N}}R_{att\_a}(n)e^{j2\pi kn/N} \quad (22)$$

$$R_{att\_a}(n) = \frac{1}{\sqrt{N}}\sum_{k=0}^{N-1} \eta_{att}s(k,n)e^{(\Theta_{R\_att}-j2\pi kn/N)}$$
$$= \eta_{att}S(n)e^{j\Theta_{R\_att}} \quad (23)$$

where

$$S(n) = \frac{1}{\sqrt{N}}\sum_{k=0}^{N-1} s(k,n)e^{-j2\pi nk/N}$$

The signal received by Bob is written as:

$$R_b(n) = \frac{1}{\sqrt{N}}\sum_{k=0}^{N-1} r_b(k,n)e^{-j2\pi nk/N}$$
$$= S_a(n)H_{a\_b}(n)e^{-j2\pi k(\varepsilon_{a\_b}-\tau_{ahead\_b})/N}$$
$$+ S_{att\_b}(n)H_{b\_att}(n)e^{-j2\pi k\varepsilon_{att\_b}/N} + Z_b(n) \quad (24)$$

where

$$S_{att\_b}(n) = \frac{1}{\sqrt{N}}\sum_{k=0}^{N-1} s_{att\_b}(k,n)e^{-j2\pi nk/N}$$
$$H_{a\_b}(n) = \sum_{l_{a\_b}=0}^{L_{a\_b}-1} h_{a\_b}(l_{a\_b})e^{-j2\pi nl_{a\_b}/N}$$
$$H_{b\_att}(n) = \sum_{l_{b\_att}=0}^{L_{b\_att}-1} h_{b\_att}(l_{b\_att})e^{-j2\pi nl_{b\_att}/N}$$
$$Z_b(n) = \sum_{k=0}^{N-1} w_b(k,n)e^{-j2\pi nk/N}$$

After Bob has received the signals, he calculates the phase offset $\Theta_{R\_b}(n)$ of the nth subcarrier from $R_b(n)$ by equations 25 and 26. η_b is the amplitude attenuation.

$$s_{R_b}(k,n) = \frac{1}{\sqrt{N}}R_b(n)e^{j2\pi kn/N} \quad (25)$$
$$R_b(n) = \eta_b S(n)e^{j\Theta_b} \quad (26)$$

As we depict above, Bob will compensate the calculated phase offset in the signal he transmits. This signal is written as equation 27 when it arrives at the attacker.

$$R_{att\_b}(n) = \frac{1}{\sqrt{N}}\sum_{k=0}^{N-1} r_{att\_b}(k,n)e^{-j2\pi nk/N}$$
$$= S_b(n)H_{b\_att}(n)e^{-j2\pi k\varepsilon_{b\_att}/N} + Z_{b\_att}(n) \quad (27)$$

where

$$S_b(n) = \frac{1}{\sqrt{N}}\sum_{k=0}^{N-1} s_b(k,n)e^{-j2\pi nk/N}$$
$$Z_{b\_att}(n) = \sum_{k=0}^{N-1} w_{b\_att}(k,n)e^{-j2\pi nk/N}$$

We can observe from this equation that the phase offset $\Theta_{R\_att'}$ consists of two kinds of phase offsets caused by the channel impulse, $H_{b\_att}(n)$ and the time offset, $e^{-j2\pi k\varepsilon_{b\_att}/N}$. Next, the attacker transmits the attack signal to Alice. The signal received by Alice is written as:

$$R_a(n) = \frac{1}{\sqrt{N}} \sum_{k=0}^{N-1} r_a(k,n) e^{-j2\pi nk/N}$$

$$= S_b(n) H_{b\_a}(n) e^{-j2\pi k(\varepsilon_{a\_b} - \tau_{ahead\_a})/N}$$

$$+ S_{att\_a}(n) H_{a\_att}(n) e^{-j2\pi k\varepsilon_{att\_a}/N} + Z_b(n) \qquad (28)$$

where

$$S_{att\_a}(n) = \frac{1}{\sqrt{N}} \sum_{k=0}^{N-1} s_{att\_a}(k,n) e^{-j2\pi nk/N}$$

$$H_{b\_a} = H_{a\_b}$$

$$Z_a(n) = \sum_{k=0}^{N-1} w_a(k,n) e^{-j2\pi nk/N}$$

Because of channel reciprocity, H_{b_a} is equal to H_{a_b}. Equation 28 tells us that there are five factors that have an impact on the phase offset Θ_{R_a} measured by the attacker from the signal transmitted by Bob: the channel impulses H_{a_b} and H_{a_att}, the time offsets ε_{a_b} and ε_{att_a} and the time difference between the legitimate signal and the attack signal, τ_{ahead_a}.

Now, it is time to extract bits from the phase offset. Also, we take a look at Alice and the attacker. Alice extracts bits of the shared key determined by $rand_{att}$, which is determined by the attacker from equation 28. If the power of the attack signal is much higher than the legitimate signal, ε_{att_a} is equal to ε_{a_att} and we ignore the negligible noise. By incorporating equations 6, 9,20 and 23 into 28, $R_a(n)$ can be expressed as equation 29, which shows that Alice can extract the correct phase offset from $\frac{A_{att}}{A_a} e^{jrand_{att}*\pi} \eta_{att} S(n)$.

$$R_a(n) = S_{att\_a}(n) H_{a\_att}(n) e^{-j2\pi k\varepsilon_{att\_a}/N}$$

$$= \frac{A_{att}}{A_a} e^{j(-\Theta_{R\_att}(n) + rand_{att}*\pi)}$$

$$\cdot S_a(n) H_{a\_att}(n) e^{-j2\pi k(\varepsilon_{att\_a} - \varepsilon_{a\_att})/N}$$

$$= \frac{A_{att}}{A_a} e^{jrand_{att}*\pi} \eta_{att} S(n) \qquad (29)$$

The attacker extracts bits of the shared key determined by Bob from equation 27. If the power of the attack signal is much higher than the legitimate signal and ε_{att_b} is equals ε_{b_att}, by incorporating equations 7, 8,24 and 26 into 27, $R_{att\_b}(n)$ can be changed into equation 30.

$$R_{att\_b}(n) = S_b(n) H_{b\_att}(n) e^{-j2\pi k\varepsilon_{b\_att}/N}$$

$$= \frac{A_b}{A_{att}} e^{j(-\Theta_{R\_b}(n) + rand_{att}*\pi)}$$

$$\cdot S_{att\_b}(n) H_{b\_att}(n) e^{-j2\pi k(\varepsilon_{b\_att} - \varepsilon_{att\_b})/N}$$

$$= \frac{A_b}{A_{att}} e^{jrand_{att}*\pi} \eta_b S(n) \qquad (30)$$

According to the above analysis, we find that there are also three factors that have an impact on the phase of the received signals: $\varepsilon_{att\_a}(\varepsilon_{att\_b})$, $\varepsilon_{a\_att}(\varepsilon_{b\_att})$ and the power ratio of the attack signal to the legitimate signal at the nth subcarrier at Alice (Bob).

IV. SIMULATION ANALYSIS

In section III-C and III-D, we analyze the factors that have an impact on the generation of the shared keys. In this section, we utilize Matlab to measure the impact of these factors on the final keys.

A. Set-up

Matlab is used to simulate the attack. All devices adopt the IEEE 802.11 OFDM protocol to communicate. Also, a multipath fading channel simulation scheme [27] is chosen to simulate the statistical wireless channel. The sampling rate is 20 MHz. The Doppler spread is 5 Hz and the root mean square delay spread is 150 ns. The power of the legitimate signals varies from 50% to 150% compared to the standard signal we choose to provide the user introduced randomness. The power of the attack signal is set to be 5, 10 and 20 dB higher than the standard signal. In 802.11a, there is a period called the guard interval that lasts 16 sample points to avoid Inter-Symbol Interference. So we specify that the time offset ε in our simulation is no larger than 16 sample points, or 0.8 ns. In our simulation, we test the key similarity rate under all possible time offset values for three kinds of power ratio. As the key is generated from both RSS quantization process and phase quantization process, we test these two processes separately.

B. RSS Quantization

As discussed in section III-C, the key generated by Alice (Bob) and the attacker from RSS quantization process is affected by three factors: $\varepsilon_{att\_a}(\varepsilon_{att\_b})$, $\varepsilon_{a\_att}(\varepsilon_{b\_att})$ and the power ratio of the received attack signals to the received legitimate signals at the nth subcarrier at Alice (Bob). We use $PowR_{att\_leg}$ to represent this power ratio. We illustrate the result of how these three factors affect the shared key in Figure 5. The unit of ε_{att_a} and ε_{a_att} is one sample point. From this figure we can observe that the key similarity between Alice and the attacker is mainly affected by $PowR_{att\_leg}$. When $PowR_{att\_leg}$ is higher, the similarity of the shared key is higher.

C. Phase Quantization

Section III-D shows the three factors that have impact on the result of phase quantization: $\varepsilon_{att\_a}(\varepsilon_{att\_b})$, $\varepsilon_{a\_att}(\varepsilon_{b\_att})$ and $PowR_{att\_leg}$. We only simulate the impact of ε_{att_a}, ε_{a_att} and $PowR_{att\_leg}$ for simplicity. We illustrate the result of how these three factors affect the shared key in Figure 6. The unit of ε_{att_a} and ε_{a_att} is one sample point. This table shows that the difference between ε_{att_a} and ε_{a_att} is the key factor that affect the similarity between the key generated by phase quantization. With the increasing of the difference between ε_{att_a} and ε_{a_att}, the key similarity varying periodically.

V. PERFORMANCE EVALUATION

To evaluate the performance of our session hijacking attack, we utilize software defined radios (USRP B200), to transmit and receive signals to acquire real CSI information as in Figure 7. The two devices are about 7 m away from each other and there are furniture and people between them. The attacker is located at one of eight different places, from A1 to A8. The power of the signals transmitted by Alice and Bob vary from 50% to 150% compared to a standard signal. The power of the attack signal at the attacker is set to be 5, 10, 20 dB
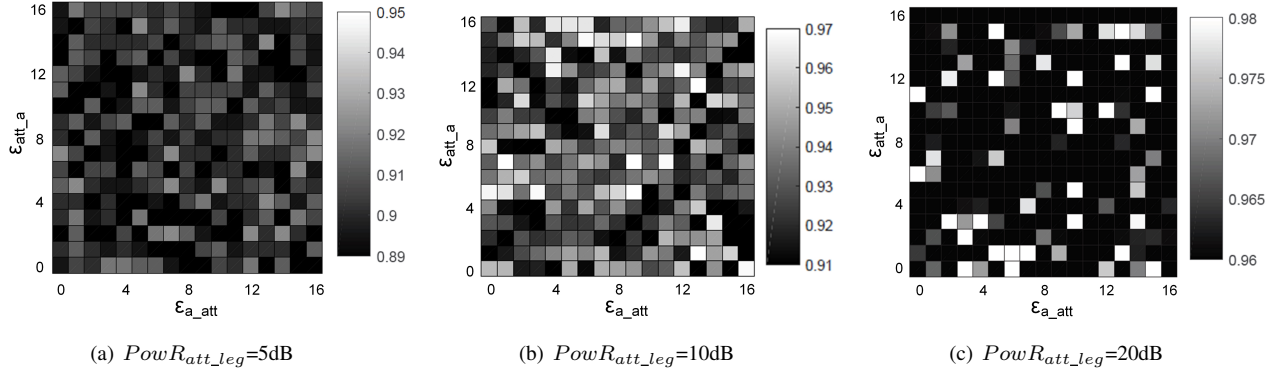
(a) $PowR_{att\_leg}$=5dB (b) $PowR_{att\_leg}$=10dB (c) $PowR_{att\_leg}$=20dB

Fig. 5: Key similarity of RSS quantization between Alice and the attacker

(a) $PowR_{att\_leg}$=5dB (b) $PowR_{att\_leg}$=10dB (c) $PowR_{att\_leg}$=20dB
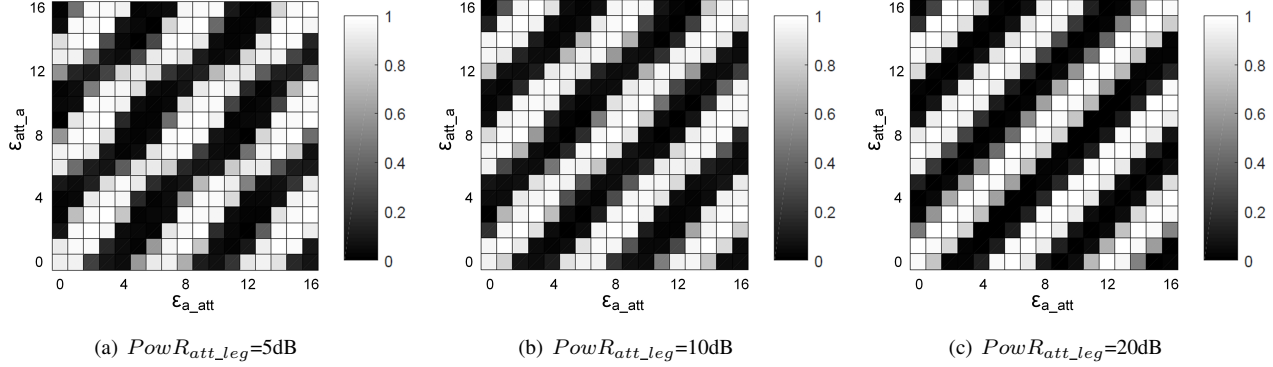
Fig. 6: Key similarity of phase quantization between Alice and the attacker

higher than the power of the standard signals in three tests. As in [14], the carrier frequency of the RF signal is 2.4 GHz and the sampling rate is 200 kHz. We implement the channel probing parts of PHY-UIR in both legitimate devices, as the main target of our attack is on this part. We also implement a timing synchronization scheme [28]. As we do not block the communication, a real-time reactive jamming technology is indispensable. In our experiment, Alice and Bob transmit signals at regular intervals. Our attacker records the arrival time of the first signals of Alice and Bob to estimate appropriate attacking time.

The result is shown in Figure 8. The Y-axis represents the similarity between the shared keys extracted at the quantization step. PowerRatio is the power ratio of the attack signal to the legitimate signal. The X-axis indicates the proximity to the legitimate device. The proximity increases from left to the right on the X-axis. We can observe that when the attacker transmits the attack signal with identical power at all eight places, the similarity of the generated keys decreases as the proximity between the attacker and legitimate device increases. This is because longer distance brings higher attenuation, which leads to lower power signals. So the similarity of the generated shared keys mainly depends on the power of the attack signal, but we also observe a sharp decline when the power ratio is 5 dB. The reason is the rapid decreasing of the timing offset when the power ratio increasing from 5 dB to 10 dB[28], which leads to more stable high similarity of the key generated by phase quantization.

When the power of the attack signal is 5 dB higher than the standard signal, and if the attacker is located further away from Alice or Bob, the key similarity rate would be too low

to achieve key agreement, due to the low power of the attack signal at the receiver. However, when the power ratio increases to 10 or 20 dB, our attack succeeds at all eight positions.

VI. ENHANCED PHY-UIR

We have shown that our session hijacking attack can break PHY-UIR. We introduce a third kind of randomness to enhance PHY-UIR against the attack. We call this enhanced scheme PHY-UIR$^+$ and describe it further in this section.

A. Workflow of PHY-UIR$^+$

PHY-UIR$^+$ prevents our attack by adding an extra step to PHY-UIR where information is simultaneously exchanged between the honest parties to ensure that the keys they agreed on are the same. In this step, PHY-UIR$^+$ specifies that honest parties should adopt binary frequency-shift keying (BFSK) technology to modulate the data. A BFSK modulation scheme is shown as

$$s(x) = \begin{cases} cos(2\pi f_0 t) & x='0' \\ cos(2\pi f_1 t) & x='1' \end{cases}$$

This scheme uses cosine waves with different frequency to represent different bits.

We show the workflow of PHY-UIR$^+$ in Figure 9. As an example, we adopt the SHA-256 hash function.

Step 1: Alice and Bob execute PHY-UIR to generate the key: K_a and K_b.

Step 2: Alice and Bob generate signals S_a and S_b separately. $S_a(S_b)$ contains message $H(K_a)(H(K_b))$, the hash value of key $K_a(K_b)$. Then they exchange the signals simultaneously.
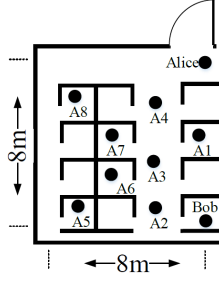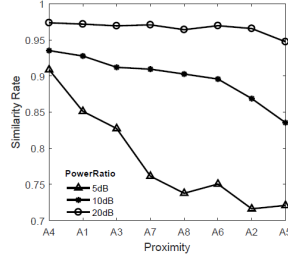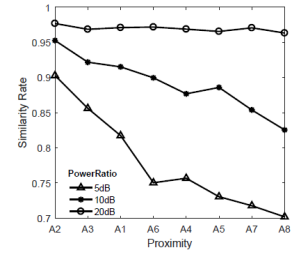
Fig. 7: Topology

(a) Key similarity rate between Alice and the attacker

(b) Key similarity rate between Bob and the attacker

Fig. 8: Similarity of the shared key

Step 3: The devices perform a noise cancellation scheme to remove the self-interference of the signal they receive. It means that the rest of the received signals at Alice and Bob are S_b and S_a respectively. Both devices recover the messages and compare the message with the hash value of their keys to check the existence of the attacker.

We can see that if K_a and K_b are the same, the results at Alice and Bob both are consistent. Inconsistency represents the existence of the attacker. Devices should transmit powerful noise to alert the other device.

If there exists an attacker, basically K_a and K_b are different from each other, then $H(K_a)$ is not equal to $H(K_b)$. The average amount of effort needed by the attacker to create two different keys amounts to a collision resistance of SHA-256, which is on average $2^{N/2} = 2^{128}$ hash attempts, which is considered infeasible to calculate. Given that a hash also exhibits a one-way property, an attacker cannot obtain additional information from the key based on the hash transmitted.
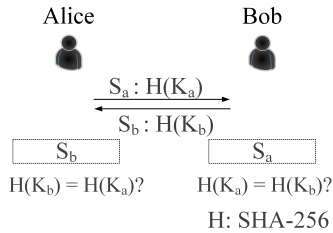

Fig. 9: Workflow of PHY-UIR$^+$

B. Performance Experiment

1) Attack Detection Performance: We conduct the experiment with the same setting in section V and Figure 7. As Alice and Bob detect the attacker by comparing the hash values of their keys, we measure the similarity between these hash values to reflect the effectiveness of PHY-UIR$^+$. The result is illustrated in Figure 10. The Y-axis refers to the similarity rate measured at Alice and Bob. We can observe from this figure that the similarity rates always locate in the range of 45% to 57% at both Alice and Bob. This means Alice and Bob can detect the attacker effectively.

2) Energy Consumption Performance: To give an indication of energy consumption of PHY-UIR$^+$ relative to PHY-UIR,
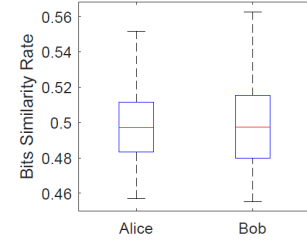
Fig. 10: Similarity rate between $H(K_a)$ and $H(K_b)$ at Alice and Bob

we implement the basic framework of PHY-UIR$^+$ and PHY-UIR in Sony Xperia e3 mobile phones. This system acts as a protocol emulator where we do the cryptographic functions and data transmissions as required, although not all signal processing functions are fully implemented, e.g. the phone cannot determine CSI. This provides a good estimation of IoT devices as Qualcomm Snapdragon processors are promoted for embedded and IoT devices, and the mobile device also allows us to measure the cost of wireless transmission (802.11), which is the main cause of power consumption. We keep running these two schemes on the phones and record the remaining power of the phones. The result is shown in Figure 11. This figure shows that PHY-UIR$^+$ uses only a little more energy than PHY-UIR which means PHY-UIR$^+$ offers improved security while being similar in term of energy efficiency.
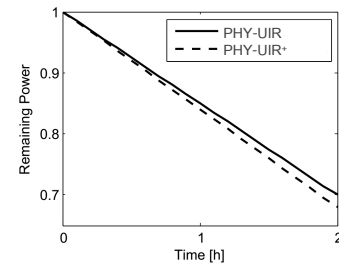
Fig. 11: Energy consumption of PHY-UIR and PHY-UIR$^+$

C. Security Analysis

In this section, we analyze the security of PHY-UIR$^+$ against the session hijacking attack. We consider three potential strategies the attacker will adopt.

1) Security Model: For the purpose of analyzing, we consider the following security model.

Definition 1 (Security Objectives): The proposed scheme has the following security requirements:

1) The scheme should detect a session hijacking attack.
2) The scheme should have no requirement of pre-shared secret information.

One of the main advantages of PHY-UIR is that it requires no pre-shared secret information to generate shared key. The solution should inherit this advantage.

Definition 2 (Adversary): The adversary is an attacker who can capture legitimate signals and send its own attack signals to overshadow target signals. The adversary has full knowledge of the design of PHY-UIR and PHY-UIR$^+$, but without unlimited computing resources.

As the protocol generates keys for each communication session, the adversary needs to launch session hijacking attacks in each communication session. Thus, security analysis considers only one protocol execution.

1) Brute-Force Search Attack: To investigate the ability of PHY-UIR$^+$ to resist brute-force search, we present Lemma 1 below.

Lemma 1: If there is no session hijacking attacker, the possibility that the eavesdropper guesses the shared key correctly at the final key exchanging step is no higher than $(\frac{1}{2})^{N_{key}}$. N_{key} is the length of the shared key.

Proof: According to [24], if the channel in a rich scattering environment is divided into small time frames, each time frame can be treated as a wide sense stationary (WSS) random process which means each time frame is uncorrelated with the other one. We also treat the OFDM signals in our paper as a WSS random process.

Each signal can extract three bits, two bits by RSS and one bit by phase. There are eight possible values and we set the possibility of these eight values as Pr_1 to Pr_8:

$$Pr_1 + Pr_2 + Pr_3 + Pr_4 + Pr_5 + Pr_6 + Pr_7 + Pr_8 = 1 \quad (31)$$

The attacker guess the value with equal possibility to choose one of the eight values. The possibility that the attacker recover the correct bits is:

$$\frac{1}{8}(Pr_1 + Pr_2 + Pr_3 + Pr_4 + Pr_5 + Pr_6 + Pr_7 + Pr_8) = \frac{1}{8} \quad (32)$$

Because the bits generated from each signal are uncorrelated with the others, the success possibility of the attacker is:

$$(\frac{1}{8})^{\frac{N_{key}}{3}} = (\frac{1}{2})^{N_{key}} \quad (33)$$

2) Signal Separation Attack: This attack tries to separate S_a and S_b from the mixed signal and then recover $H(K_a)$ and $H(K_b)$ with the purpose of recovering K_a and K_b. As a hash function is a one-way function, it is hard to deduce the key from its hash value.

3) Smart Session Hijacking Attack: Let's consider a smart session hijacking attack where the attacker tries to overshadow the extra step that PHY-UIR$^+$ adds after it launches the session attack. We illustrate this attack in Figure 12.

The attacker transmits signal with malicious data Att, which has the same length with $H(K_a)$ and $H(K_b)$. Now we are going to prove the effectiveness of our method against the smart session hijacking attack.
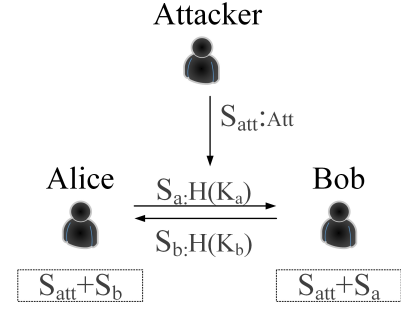

Fig. 12: Smart session hijacking attack

Lemma 2: When the attacker and Alice(Bob) transmit the same bit at the extra step in PHY-UIR$^+$, Bob(Alice) receives the same bit.

Proof: When Alice and the attacker transmit the same bit, the signal Bob receives is expressed as equation 34. As they sent the same bit, both signals have identical frequency f, but they have different amplitude and phase shift.

$$\begin{aligned} R_b &= S_{att} + S_a \\ &= \alpha_{att}cos(2\pi ft + \theta_{att}) + \alpha_a cos(2\pi ft + \theta_a) \end{aligned} \quad (34)$$

where α_{att} and α_a are the amplitude of signals from the attacker and Alice; θ_{att} and θ_a are the phase offset of signals from the attacker and Alice.

Equation 34 can change into equation 35 [29].

$$R_b = \alpha cos(2\pi ft + \theta) \quad (35)$$

where

$$\begin{aligned} \alpha &= \sqrt{\alpha_{att}^2 + \alpha_a^2 + 2\alpha_{att}\alpha_a cos\Delta\theta} \\ \Delta\theta &= \theta_{att} - \theta_a \\ tan\theta &= \frac{\alpha_{att}sin\theta_{att} + \alpha_a sin\theta_a}{\alpha_{att}cos\theta_{att} + \alpha_a cos\theta_a} \end{aligned}$$

We can observe from equation 35 that R_b has the same frequency as S_{att} and S_a, which means Bob receives the same bit that is transmitted by Alice and the attacker. Similarly we can prove that when Bob and the attacker transmit the same bit, Alice receives the same bit.

Lemma 3: When a smart session hijacking attack has been launched successfully and there are difference between hash values, $H(K_a)$ and $H(K_b)$, generated by two legitimate participants, PHY-UIR$^+$ can always detect the attack.

Proof: First, we specify the operation that legitimate participants use to recover data from received signal as $\bigotimes$. When the attacker and Alice(Bob) transmit different bits, the final bit recovered by Bob(Alice) can be either of those two different bits, usually the one with higher energy. Adding Lemma 2, we can write $\bigotimes$ as

$$A \bigotimes B = \begin{cases} A & \text{A=B} \\ A \text{ or } B & \text{A} \neq \text{B} \end{cases}$$

As we set a precondition that $H(K_a)$ and $H(K_b)$ have differences, we only consider the differences. Analyzing at bit level, one bit transmitted by Alice is b and the corresponding bit transmitted by Bob is $\neg b$. This is because BFSK transmits binary bits. The result at Alice is $Att \bigotimes \neg b$ while the result at Bob is $Att \bigotimes b$. According to the workflow in Figure 9,

the adversary needs to satisfy the following two equations to avoid being detected:

$$Att \bigotimes \neg b = b \qquad (36)$$

$$Att \bigotimes b = \neg b \qquad (37)$$

If the adversary wants to satisfy equation 36, Att should be equal to b which will make equation 37 false, vice versa. This means either Alice or Bob will detect the attack. So Lemma 3 is proved.

VII. RELATED WORK

The signal manipulation attack is a well-known attack in wireless networks. There are three kinds of signal manipulation attacks [30]: the disruptive jamming attack, the manipulative jamming attack and the channel manipulation attack. The disruptive jamming attack is the blocking attack we discussed in the introduction. The manipulative jamming attack injects a signal to manipulate the generated key while the channel manipulation attack controls the channel to infer the key.

There are five practical key manipulation attacks: three for manipulative jamming [14], [13], [31] and two for the channel manipulation [32], [33] by now.

In [14], an attack which aims at forcing both victims (devices) to generate the same shared key that is known to the attacker by injecting signals, is proposed. But it needs to have knowledge of the CSI of the channels between victims to the attacker. Another similar method [13] also wants to control the key but it requires waiting for an opportunity to inject the attack signals, which leads to a low key manipulation rate. The idea of [31] is relaying legitimate signals with higher power to overshadow the legitimate signal. Hence, there exists non-trivial probability that the receiver will extract they key from the information of the channel between the attacker and the receiver.

Instead of injecting signals to manipulate the key, [32] proposed a novel method to control the channel by controlling the movements of the intermediate objects or the devices themselves. Then the CSI of the channel will change in a predictable way, which can be used to infer the shared key extracted from the controlled CSI. However, introducing an additional random variable that is unknown to the attacker [33] can stop this attack.

Our proposed attack has fewer constraints compared to [14], [13], [33]. Our attack has no need for prior knowledge of the CSI of the channels or wait for a specific moment to attack. Another advantage of our attack is that it can break the proposed security schemes that protects against the attacks in [14], [32], [33].

VIII. CONCLUSION

In this paper we introduced a novel active attack, called a session hijacking attack, against an existing physical layer-key generation scheme with user introduced randomness (PHY-UIR). Instead of attempting to reveal the key generated by legitimate participants, our attack forces participants to run the key generation scheme with the attacker by injecting attack signals.

We analyze the impact of the session hijacking attack on two quantization schemes, RSS and signal phase, used in PHY-UIR. Then we study the effectiveness of our attack on these two schemes by simulation. Results show that the power ratio between the attack signal and the legitimate signal at Alice (Bob) is the crucial factor to success or failure of the RSS quantization attack, while for the signal phase quantization attack the key factor is the time offset. Experiments were also conducted to validate the effectiveness of our attack and the findings of the simulation.

In the end, we propose PHY-UIR$^+$ to detect our session hijacking attack by adding an extra step for both devices exchanging information about the shared keys at the same time. Experiments have shown that PHY-UIR$^+$ can detect the proposed session hijacking attack effectively and efficiently. Finally, we analyze the security of this method and prove the resistance to brute-force attacks and enhanced session hijacking attacks.

REFERENCES

[1] D. Abbasinezhad-Mood and M. Nikooghadam, "Efficient anonymous password-authenticated key exchange protocol to read isolated smart meters by utilization of extended Chebyshev chaotic maps," *IEEE Transactions on Industrial Informatics*, 2018.

[2] R. Arul, G. Raja, A. K. Bashir, J. Chaudhry, and A. Ali, "A console GRID leveraged authentication and key agreement mechanism for LTE/SAE," *IEEE Transactions on Industrial Informatics*, 2018.

[3] M. Wang and Z. Yan, "Privacy-preserving authentication and key agreement protocols for D2D group communications," *IEEE Transactions on Industrial Informatics*, 2017.

[4] N. Saxena and S. Grijalva, "Dynamic secrets and secret keys based scheme for securing last mile smart grid wireless communication," *IEEE Transactions on Industrial Informatics*, vol. 13, no. 3, pp. 1482–1491, 2017.

[5] Y. Qiu and M. Ma, "A mutual authentication and key establishment scheme for M2M communication in 6LoWPAN networks," *IEEE Transactions on Industrial Informatics*, vol. 12, no. 6, pp. 2074–2085, 2016.

[6] S. Mathur, W. Trappe, N. Mandayam, C. Ye, and A. Reznik, "Radio-telepathy: extracting a secret key from an unauthenticated wireless channel," in *Proceedings of the 14th ACM international conference on Mobile computing and networking*. ACM, 2008, pp. 128–139.

[7] C. Ye, S. Mathur, A. Reznik, Y. Shah, W. Trappe, and N. B. Mandayam, "Information-theoretically secret key generation for fading wireless channels," *IEEE Transactions on Information Forensics and Security*, vol. 5, no. 2, pp. 240–254, 2010.

[8] R. Ahlswede and I. Csiszár, "Common randomness in information theory and cryptography. i. secret sharing," *IEEE Transactions on Information Theory*, vol. 39, no. 4, pp. 1121–1132, 1993.

[9] Y. Wu, A. Khisti, C. Xiao, G. Caire, K.-K. Wong, and X. Gao, "A survey of physical layer security techniques for 5g wireless networks and challenges ahead," *IEEE Journal on Selected Areas in Communications*, vol. 36, no. 4, pp. 679–695, 2018.

[10] J. Zhang, T. Q. Duong, A. Marshall, and R. Woods, "Key generation from wireless channels: A review," *IEEE Access*, vol. 4, pp. 614–626, 2016.

[11] F. Pan, Z. Pang, M. Luvisotto, M. Xiao, and H. Wen, "Physical-layer security for industrial wireless control systems: Basics and future directions," *IEEE Industrial Electronics Magazine*, vol. 12, no. 4, pp. 18–27, Dec 2018.

[12] S. M. MirhoseiniNejad, A. Rahmanpour, and S. M. Razavizadeh, "Phase jamming attack: A practical attack on physical layer-based key derivation," in *2018 15th International ISC (Iranian Society of Cryptology) Conference on Information Security and Cryptology (ISCISC)*. IEEE, 2018, pp. 1–4.

[13] S. Eberz, M. Strohmeier, M. Wilhelm, and I. Martinovic, "A practical man-in-the-middle attack on signal-based key generation protocols," in *European Symposium on Research in Computer Security*. Springer, 2012, pp. 235–252.

[14] R. Jin and K. Zeng, "Physical layer key agreement under signal injection attacks," in *Communications and Network Security (CNS), 2015 IEEE Conference on*. IEEE, 2015, pp. 254–262.

[15] V. Brik, S. Banerjee, M. Gruteser, and S. Oh, "Wireless device identification with radiometric signatures," in *Proceedings of the 14th ACM international conference on Mobile computing and networking*. ACM, 2008, pp. 116–127.

[16] B. Danev and S. Capkun, "Transient-based identification of wireless sensor nodes," in *Proceedings of the 2009 International Conference on Information Processing in Sensor Networks*. IEEE Computer Society, 2009, pp. 25–36.

[17] M. Demirbas and Y. Song, "An rssi-based scheme for sybil attack detection in wireless sensor networks," in *Proceedings of the 2006 International Symposium on on World of Wireless, Mobile and Multimedia Networks*. IEEE Computer Society, 2006, pp. 564–570.

[18] N. Patwari and S. K. Kasera, "Robust location distinction using temporal link signatures," in *Proceedings of the 13th annual ACM international conference on Mobile computing and networking*. ACM, 2007, pp. 111–122.

[19] B. Danev, H. Luecken, S. Capkun, and K. El Defrawy, "Attacks on physical-layer identification," in *Proceedings of the third ACM conference on Wireless network security*. ACM, 2010, pp. 89–98.

[20] Y.-C. Tung, K. G. Shin, and K.-H. Kim, "Analog man-in-the-middle attack against link-based packet source identification," in *Proceedings of the 17th ACM International Symposium on Mobile Ad Hoc Networking and Computing*. ACM, 2016, pp. 331–340.

[21] Q. Hu and G. P. Hancke, "A session hijacking attack on physical layer key generation agreement," in *2017 IEEE International Conference on Industrial Technology (ICIT)*. IEEE, 2017, pp. 1418–1423.

[22] K. Zeng, D. Wu, A. Chan, and P. Mohapatra, "Exploiting multiple-antenna diversity for shared secret key generation in wireless networks," in *INFOCOM, 2010 Proceedings IEEE*. IEEE, 2010, pp. 1–9.

[23] H. Minn, V. K. Bhargava, and K. B. Letaief, "A robust timing and frequency synchronization for OFDM systems," *IEEE Transactions on Wireless Communications*, vol. 2, no. 4, pp. 822–839, 2003.

[24] A. F. Molisch, K. Balakrishnan, C.-C. Chong, S. Emami, A. Fort, J. Karedal, J. Kunisch, H. Schantz, U. Schuster, and K. Siwiak, "Ieee 802.15. 4a channel model-final report," *IEEE P802*, vol. 15, no. 04, p. 0662, 2004.

[25] Q. Wang, H. Su, K. Ren, and K. Kim, "Fast and scalable secret key generation exploiting channel phase randomness in wireless networks," in *INFOCOM, 2011 Proceedings IEEE*. IEEE, 2011, pp. 1422–1430.

[26] K.-C. Hsu, K. C.-J. Lin, and H.-Y. Wei, "Full-duplex delay-and-forward relaying," in *Proceedings of the 17th ACM International Symposium on Mobile Ad Hoc Networking and Computing*. ACM, 2016, pp. 221–230.

[27] C.-D. Iskander and H.-T. Multisystems, "A MATLAB-based object-oriented approach to multipath fading channel simulation," *Hi-Tek Multisystems*, vol. 21, 2008.

[28] A. A. Nasir, S. Durrani, and R. A. Kennedy, "Performance of coarse and fine timing synchronization in OFDM receivers," in *2010 2nd International Conference on Future Computer and Communication*, vol. 2. IEEE, 2010, pp. V2–412.

[29] W. Benenson, J. W. Harris, H. Stöcker, and H. Lutz, *Handbook of physics*. Springer Science & Business Media, 2006.

[30] K. Zeng, "Physical layer key generation in wireless networks: challenges and opportunities," *IEEE Communications Magazine*, vol. 53, no. 6, pp. 33–39, 2015.

[31] Y. Qiao, K. Srinivasan, and A. Arora, "Channel spoofer: Defeating channel variability and unpredictability," in *Proceedings of the 13th International Conference on emerging Networking EXperiments and Technologies*. ACM, 2017, pp. 402–413.

[32] S. Jana, S. N. Premnath, M. Clark, S. K. Kasera, N. Patwari, and S. V. Krishnamurthy, "On the effectiveness of secret key extraction from wireless signal strength in real environments," in *Proceedings of the 15th annual international conference on Mobile computing and networking*. ACM, 2009, pp. 321–332.

[33] R. Jin and K. Zeng, "Manipulative attack against physical layer key agreement and countermeasure," *IEEE Transactions on Dependable and Secure Computing*, 2019.
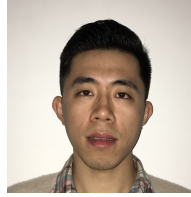
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